

Real Time Parallel Architecture in VLSI using Microcontroller

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Abstract

Abstract- This paper proposes a novel architecture for high speed arithmetic by a “multiplier-and-accumulator (MAC)”. A hybrid type of CSA is developed by the combination of multiplication and accumulation. The largest delay accumulator in MAC that are merged into CSA and evaluation of overall performance. The CSA trees proposed here use a “1’s-complement-based radix-2 modified Booth’s algorithm (MBA)” and the array is modified for extending sign for increasing operand’s bit density. The carry is propagated to LSBs of partial products by CSA and LSB is generated for decreasing final adder’s input bits. The intermediary results are accumulated in proposed MAC resulting in carry and sum bits instead of final adder output, which makes optimization of pipeline scheme for improving performance. The synthesis of architecture proposed here was done using standard CMOS library of 90,130,180 and 250 nm. The results are analysed based on experimental and theoretical estimation such as delay, pipelining scheme and hardware resources. The delay modelling use Sakurai’s alpha power law. The MAC proposed here is superior in properties in comparison with standard design and it has twice performance at same clock frequency as previous research and it could be applied to high performance requirements.

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I. INTRODUCTION

The demands in real time processing of signals such as audio signals, large capacity of data processing and video/image processing are increasing with an advancement in communication and multimedia systems. The essential elements of DSP are multiplier and “multiplier-and accumulator (MAC)”[1]. They perform inner products, filtering and convolution. Mostly non-linear functions are utilized by methods of digital signal processing such as “Discrete Cosine Transform (DCT)” [2] and “Discrete Wavelet Transform (DWT)”[3]. Because of repetitive addition and multiplication, the execution of speed proceedings and entire calculation are determined by the addition and

multiplication speeds. The critical path in a digital system is calculated by a multiplier as it requires largest delay among basic blocks of operation. A “modified radix-4 Booth’s algorithm (MBA)” [4] is used in applications of high speed multiplication. However the problems associated with long critical path cannot be solved by this [5]. A multiplier comprise of full adders (FAs), Booth’s algorithm [3] and a Wallace tree [5] multiplier inspite of an array of full adders. Thus, there are three parts of this multiplier: a booth encoder, a Wallace tree and a final adder [6][7]. The Wallace tree adds encoder’s partial products in a parallel manner and its time of operation is in direct proportion to the number of inputs. The fact that the number of outputs are reduced when

number of 1's are counted among inputs. The number of outputs are reduced in each step of pipeline by using (3:2) and (7:3) counters. The speed of a multiplier can be effectively increased by reducing partial products as addition series are proceeded for partial products. MBA algorithm is used for reducing the steps for calculation of partial products. The speed of addition of partial products is increased by a Wallace tree. The research for various parallel architectures has been done for increasing the speed of MBA algorithm [8][9]. Among those research work, architectures based on "Baugh-Wooley algorithm (BWA)" was found to be best one.

II. VLSI ARCHITECTURE

The progress of regular machines has exploited locality and concurrency by the anticipated scaling of VLSI and it was programmable. The table 1 has two columns 2 and 3 which shows a 1000 fold increase in grid numbers in an expected period of twenty years and hence the fabrication of devices on a chip economically. The device count can be increased for performance by using clear concurrency (parallelism). There is a requirement of locality as the bandwidth of wire at module's periphery was scaling as a device count's square root which is comparatively slower than two third of power needed by Rent's rule. It was seen in 1979 that wires limit performance, power and area of various modules. The programmability and regularity are motivated by issues of complexity in design[10].

It is easier to design simple processing and identical nodes in an array rather than a complex "multi-million transistor processor. The mounting of design costs is beamoritized for a wide variety of applications by a programmable design. The design can be understood by an efficient network that connects together processor arrays [Dally92, DYN97][3]. Mechanisms allowing quick synchronization and communication of processors have been developed over these networks [LDK+98]. The implementation of coherent and efficient shared memory systems is understood [ASHH88]. Also the parallel machines have been programmed by several methods[11]. The

technology is demonstrated by constructing research machines and parallel software research platform. A general purpose network such as 3-D torus is able to outperform a network having a topology which matches with the problem of interest such as tree for divide and conquer problems. Thus, it is advisable to provide general purpose mechanism set instead of a specialized machine for a single computation model. At the success of high end, there is a little impact of parallel VLSI architectures on mainstream computer industry. A few 10s of processors are contained in departmental stores and desktop machines are uniprocessors. The microprocessor chips are capable of holding 1000 of 8086, still area is required for implementing a single processor. The three major reasons for course of events is a considerable opportunity for applying additional grids for improving performance of sequential processors, compatibility of software for favoured sequential machines and mechanisms used in parallel machines that inspired a coarse granularity of software and hardware. In 1979, the difference in performance is larger than 100's factor between best and high end CPU such as used in Cray 1 or IBM 370. The difference in gate delay between MOS technology and bipolar resulted in a small difference of a factor 3. The difference existed because of increased count of gates that pipelined the execution aggressively and exploited parallelism. Most of the unconventional features were incorporated by microprocessors which closed the gap between 1979 and 1999. The table 1 illustrates an increase in the clock frequency by 80: factor 20 is because of delay in gate and 4 is because of reduction in gate numbers per clock. The number of clocks per instruction are reduced by a factor 12.5 from 10.

III. MIMO SYSTEM MODEL

MIMO communication systems require multiple antennas at the transmitter and receiver sections. Figure 1 illustrates spatial multiplexing and transmit antennas that are capable of transmitting M symbols where N receive antennas are present for receiving every symbol. Thus, multiple

channel path is created because all the transmitted symbols are received by receiving antennas. On combining these paths, a matrix of channel elements is formed. N channel paths are created by each symbol and N antennas receive these symbols. Because of simultaneous transmission of M symbols, an NxM matrix is formed.

$$\mathbf{s} = (s_1, s_2, \dots, s_M)^T \text{ And } \mathbf{r} = (r_1, r_2, \dots, r_M)^T$$

Here, s denotes the transmitted symbol vector and r denotes the received signal and H is the channel matrix of NxM between received and transmit antenna array. The independent AWGN is denoted by y and noise vector is distributed identically.

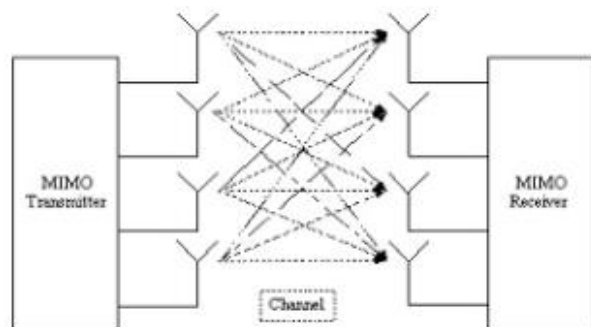


Fig. 1 System model of MIMO

A. Square Root Algorithm for V-BLAST

VBLAST includes successive cancellation and nulling for detecting transmitted symbols. The symbol that has highest value of detection of “Signal to Noise ratio (SNR)” is determined by inverting and reordering channel matrix. This relates to an inverted channel matrix row that has smallest Euclidean distance. After a signal is detected, it is subtracted from a symbol vector which is received. The column of H matrix related to it is zeroed down and process is continued to repeat with a deflated channel matrix until the detection of all symbols is done. This research paper uses MMSE for channel inversion. A generic matrix H has its pseudo inverse is:

$$H^+ = (H^*H)^{-1}H^* = R^{-1}Q^* \quad (3)$$

Either QR decomposition or “singular value decomposition (SVD)” is used for computing

pseudo inverse. The MMSE-VBLAST is computed by “square root algorithm” [3] and augmented channel matrix’s QR decomposition.

$$H N \times M \text{ or } I M \times M = QR = Q a N \times M \times R M \times M \quad (4)$$

c. Irrelevant entries are denoted by x. The channel matrix is first decomposed into

$$QR \text{ or } a + ja1$$

And then $P1/2 = R^{-1}$ is computed. The repeated computation of pseudo inverse can be neglected after computing Qa and P1/2.

IV. CORDIC

CORDIC is accomplishes rotation in hardware efficiently by implementing the rotation equations: $x' = \cos\theta x - y \tan\theta$, $y' = \cos\theta y + x \tan\theta$ (7) Angles are selected in such a way that $\tan\theta = 2^{-i}$ (8)

A right shift follows multiplication in this case. Several processing elements of CORDIC are together used which can be rotated by combining allowed angles by an arbitrary angle: $\theta = \tan^{-1} 2^{-i}$ (9)

The terms are constant for rotating for fixed iteration numbers. Up to 15 iterations, a constant value of scaling is seen [12]. This design requires a CORDIC for rotating a vector to nulling axis and then angles which are related to following vectors are rotated to same angle. These operation modes are called rotation and vectoring. The constraint on angles was used for implementing equations of rotation this design of CORDIC [9] in such a way it results in nulls. It is also required to design a CORDIC capable of operating in rotation and vectoring mode. The equations are implemented by using adders and shifters for doing bulk of work. Two input vectors are received by every processing element and their sign is found [13]. The decision about up or down rotation is made based on signs by using SIMULINK. It enables the creation and use of block diagrams of high level which is utilized for emulation, hardware description and simulation. This design used Xilinx block set. The

architecture of “pseudo inverse” is designed by using a block set of Xilinx[14].

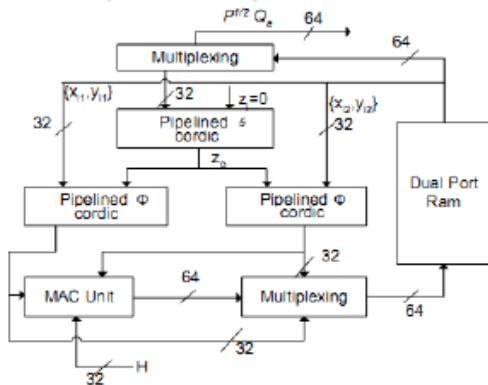


Fig.2 Architecture for pseudo inverse in VLSI

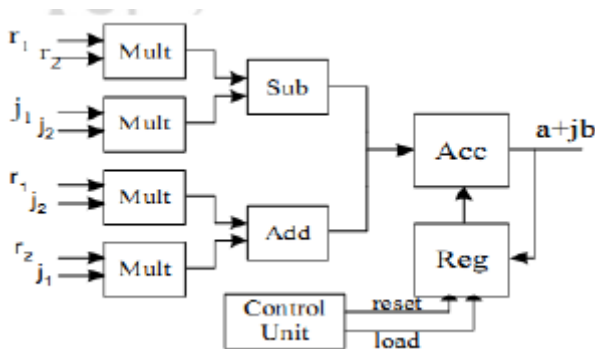


Fig. 3 MAC module

V. RESULTS

Extensive simulation of every block is done during simulation. The algorithm is performed for testing that block in Xilinx ISE, system generator and MATLAB for obtaining expected values that are given in certain data. The inputs similar to algorithm is given to blocks is given in MATLAB, running the simulation and reviewing blocks[15]. The Pseudo Inverse module consumed a total power of about 239mW.

Supply Summary		Total	Dynamic	Quiescent
Source	Voltage	Current (A)	Current (A)	Current (A)
Vccint	1.200	0.110	0.093	0.017
Vccaux	2.500	0.016	0.001	0.015
Vcco25	2.500	0.027	0.025	0.002
Supply Power (W)		Total	Dynamic	Quiescent
		0.239	0.178	0.061

Fig. 4 PINV module power analysis

A	B	C	D	E	F	G	H
Device		On-Chip	Power (W)	Used	Available	Utilization (%)	
Family	Spartan3	Clocks	0.009	1	
Part	xc3s400	Logic	0.037	1911	7168	26.7	
Package	ft256	Signals	0.061	3136	
Grade	Commercial	I/Os	0.068	57	173	32.9	
Process	Typical	BRAMs	0.002	1	16	6.3	
Speed Grade	-4	MULTs	0.001	4	16	25.0	
		Leakage	0.061				
		Total	0.239				
Environmental		Thermal Properties					
Ambient Temp (C)	25.0	Use custom TJA?	No	Effective TJA	Max Ambient	Junction Temp	
Custom TJA (C/W)	NA	Airflow (LFM)	0	(C/W)	(C)	(C)	
				27.9	78.3	31.7	
Characterization							
PRODUCTION	v1.2.06-25-09						

Fig. 5 Device family used for Pseudo Inverse module

Table 1 Summary of device utilization

Device Utilization Summary				
Logic Utilization	Used	Available	Utilization	Note(s)
Number of Slice Flip Flops	1,703	7,168	23%	
Number of 4 input LUTs	1,714	7,168	23%	
Number of occupied Slices	1,184	3,584	33%	
Number of Slices containing only related logic	1,184	1,184	100%	
Number of Slices containing unrelated logic	0	1,184	0%	
Total Number of 4 input LUTs	1,911	7,168	26%	
Number used as logic	1,639			
Number used as a route-thru	197			
Number used as Shift registers	75			
Number of bonded IOBs	57	173	32%	
Number of RAMB16s	1	16	6%	
Number of MULT18X18s	4	16	25%	
Number of BUFMGXs	1	8	12%	
Average Fanout of Non-Clock Nets	1.94			

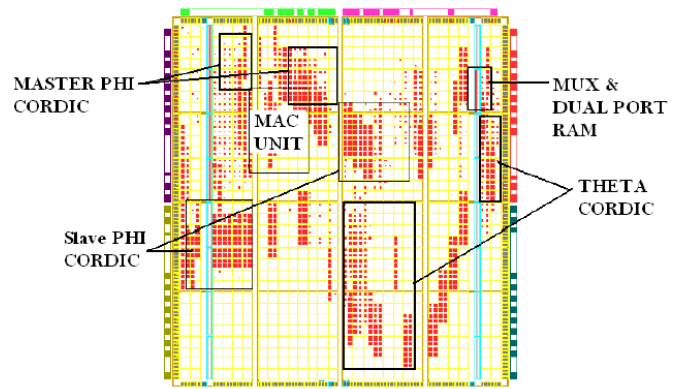


Fig. 6 Layout of pseudo inverse module

VI. CONCLUSION

A VLSI architecture which has single processor instead of large number processor based QR triangular array is used in detection of V-BLAST. The tradeoff is considered between performance and hardware complexity and then square root algorithm presents its quantization scheme for V-BLAST detection. Special XILINX block sets were used for implementing proposed architecture in SIMULINK. The information stream is subsequently decoded in V-BLAST architecture

and SIC is easily performed after finding $p1/2$. The future work involves addressing design and implementation of different modules of NULL and SORT for area utilization and power.

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